

# Coprime factorizations and stabilizability of infinite-dimensional linear systems

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# Main Theorem

The following are equivalent for a holomorphic function  $P$ :

- (i)  $P$  has a dynamic **stabilizing controller**.
- (ii)  $P$  has a right **coprime factorization**. [Smith89] [M05d]
- (iii)  $P$  has a **stabilizable and detectable realization**. [Staffans98] [CurOpm05]  
[M05c]

We work in discrete time, but essentially the same results hold in continuous time too. Part of the results are new even in the scalar-valued case.

As corollaries, one obtains analogous results for exponential (power) stabilization.

# Notation

$U, X, Y$ : complex Hilbert spaces of arbitrary dimensions.

$\mathbb{D}$ : the unit disc  $\{z \in \mathbb{C} \mid |z| < 1\}$ .

$\mathcal{B}(U, Y)$ : bounded linear maps  $U \rightarrow Y$ .

$H^\infty(U, Y)$ : the set of bounded holomorphic functions  $\mathbb{D} \rightarrow \mathcal{B}(U, Y)$ .

$I$ : the identity operator, e.g.,  $I = I_U \in \mathcal{B}(U, U)$ , or the corresponding constant function, e.g.,  $I = I_U \in H^\infty(U, U)$ .

**proper function** = holomorphic (operator-valued) function defined near the origin;

**strictly proper** =  $P$  is proper and  $P(0) = 0$ ;

**stable** =  $H^\infty$  (a restriction of a  $H^\infty$  function is identified with the  $H^\infty$  function).

Motivation:  $P \in H^\infty(U, Y) \implies P$  is bounded (stable) multiplication operator  $H^2(U) \rightarrow H^2(Y)$ .

# Dynamic (output-feedback) stabilization

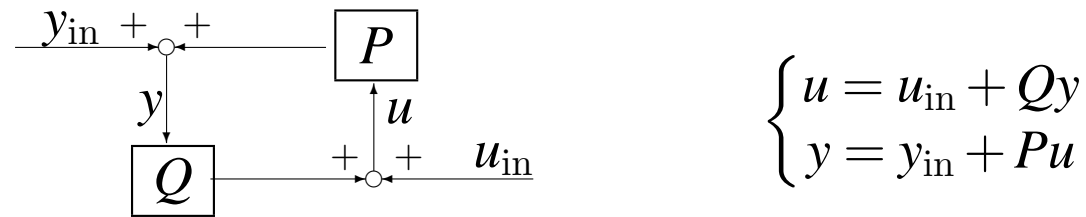


Figure 1: Controller  $Q$  for the transfer function  $P$

**stabilizing controller**  $= \begin{bmatrix} u_{\text{in}} \\ y_{\text{in}} \end{bmatrix} \mapsto \begin{bmatrix} u \\ y \end{bmatrix}$  is stable ( $H^\infty$ ).

A proper  $\mathcal{B}(Y, U)$ -valued function  $Q$  is called a (dynamic output feedback) **proper stabilizing controller** for a proper  $\mathcal{B}(U, Y)$ -valued function  $P$  if the “input-to-error” map  $E : \begin{bmatrix} u_{\text{in}} \\ y_{\text{in}} \end{bmatrix} \mapsto \begin{bmatrix} u \\ y \end{bmatrix}$  in Figure 1 is stable ( $E \in H^\infty$ ). The map  $E$  is obviously given by

$$E := \begin{bmatrix} I & -Q \\ -P & I \end{bmatrix}^{-1} = \begin{bmatrix} (I - QP)^{-1} & Q(I - PQ)^{-1} \\ P(I - QP)^{-1} & (I - PQ)^{-1} \end{bmatrix}. \quad (1)$$

(Observe that then  $P$  is also a proper stabilizing controller for  $Q$ .)

## Right coprime

The following are equivalent for a proper holomorphic function  $P$ :

- (i)  $P$  has a proper **stabilizing controller**  $Q$  (i.e.,  $\begin{bmatrix} I & -Q \\ -P & I \end{bmatrix}^{-1} \in H^\infty$ ).
- (ii)  $P$  has a right **coprime factorization**.
- (iii)  $P$  has a **stabilizable and detectable realization**.

Two functions  $M, N \in H^\infty$  are called (Bézout) **right coprime** if  $\begin{bmatrix} M \\ N \end{bmatrix}$  is left-invertible in  $H^\infty$ , i.e., if there exist  $\tilde{X}, \tilde{Y} \in H^\infty$  satisfying the *Bézout identity*

$$\tilde{X}M - \tilde{Y}N \equiv I \quad (\text{on } \mathbb{D}) . \quad (2)$$

We call the factorization  $P = NM^{-1}$  a **right coprime factorization** of  $P$  if  $N \in H^\infty(\mathbb{U}, \mathbb{Y})$  and  $M \in H^\infty(\mathbb{U})$  are right coprime,  $M(0)$  is invertible and  $P = NM^{-1}$ .

Then  $Q = \tilde{X}^{-1}\tilde{Y}$  is a stabilizing controller for  $P$  (if  $\tilde{X}^{-1}$  exists).

## All stabilizing controllers

Let  $P$  be  $\mathcal{B}(U, Y)$ -valued and have a right coprime factorization  $P = NM^{-1}$ . Then  $\begin{bmatrix} M \\ N \end{bmatrix} \in H^\infty(U, U \times Y)$  can be extended to an invertible element of  $H^\infty(U \times Y)$ , say  $\begin{bmatrix} M & Y \\ N & X \end{bmatrix}$ . (This is called a *doubly coprime factorization* of  $P$ .) [Tolokonnikov81] [Treil04] [M05d]

All stabilizing controllers for  $P$  are given by the **Youla(-Bongiorno) parameterization**

$$Q = (Y + MV)(X + NV)^{-1} \quad (3)$$

where  $V \in H^\infty(Y, U)$  is arbitrary (the controller is proper iff  $(X + NV)^{-1}$  is proper). [CuWeWe01] [M05d]

If  $P$  is strictly proper ( $P(0) = 0$ ), then all these controllers are proper.

## Matrix-valued case

Let  $P$  be a proper  $\mathbb{C}^{n \times m}$ -valued function. Then also the following are equivalent to the existence of a proper stabilizing controller:

- (i\*)  $P$  has a **stable** ( $Q \in H^\infty(\mathbb{C}^n, \mathbb{C}^m)$ ) stabilizing controller. [Treil92] [Quadrat04]
- (ii\*)  $P = NM^{-1}$ , where  $N, M \in H^\infty$ ,  $N^*N + M^*M \geq \varepsilon I$  on  $\mathbb{D}$ ,  $\varepsilon > 0$  and  $\det M \neq 0$ .  
[Carleson62] [Fuhrman68]

(The corona condition in (ii') is not sufficient for coprimeness in the operator-valued case [Treil89]. It is not known whether (i') is necessary in general.)

## Controllers with internal loop

Also the following is equivalent to the existence of a proper stabilizing controller of  $P$ :

(i'')  $P$  has a stabilizing controller with **internal loop**. [CuWeWe01] [M05d]

We call  $R$  a **stabilizing controller with internal loop** for  $P$  if  $R = \begin{bmatrix} R_{11} & R_{12} \\ R_{21} & R_{22} \end{bmatrix}$  is a proper  $\mathcal{B}(Y \times \Xi, U \times \Xi)$ -valued function for some Hilbert space  $\Xi$  and the combined map  $\begin{bmatrix} u_{\text{in}} \\ y_{\text{in}} \\ \xi_{\text{in}} \end{bmatrix} \mapsto \begin{bmatrix} u \\ y \\ \xi \end{bmatrix}$  in Figure 2 becomes stable ( $H^\infty$ ).

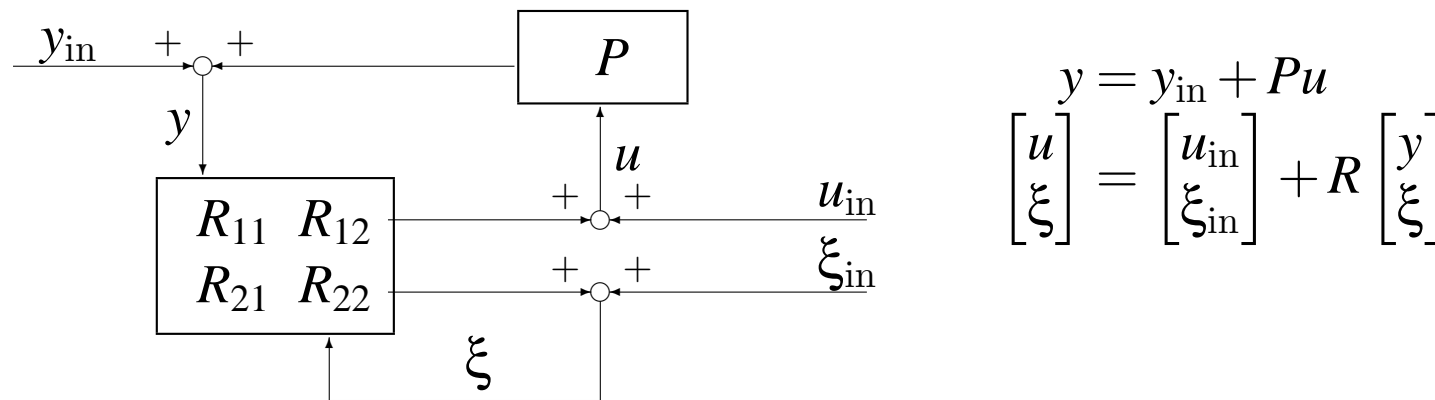


Figure 2: Controller  $R$  with internal loop for  $P$

If  $I - R_{22}(0)$  is invertible, then  $R$  corresponds to the proper stabilizing controller  $Q = R_{11} + R_{12}(I - R_{22})^{-1}R_{21}$ .



## Main Theorem (ver. 3)

The following are equivalent for a proper function  $P$ :

- (i)  $P$  has a proper **stabilizing controller**  $Q$  (i.e.,  $\begin{bmatrix} I & -Q \\ -P & I \end{bmatrix}^{-1} \in \mathbf{H}^\infty$ ).
- (i')  $P$  has a **strictly proper** stabilizing controller.
- (i'')  $P$  has a stabilizing controller with **internal loop**.
- (ii)  $P$  has a **right coprime** factorization  $P = NM^{-1}$ .
- (ii')  $P$  has a **left coprime** factorization  $P = \tilde{M}^{-1}\tilde{N}$ .
- (ii'')  $P$  has a **doubly coprime** factorization  $P = NM^{-1}$ ,  $\begin{bmatrix} M & Y \\ N & X \end{bmatrix}, \begin{bmatrix} M & Y \\ N & X \end{bmatrix}^{-1} \in \mathbf{H}^\infty(\mathbf{U} \times \mathbf{Y})$ .
- (iii)  $P$  has a **stabilizable and detectable realization**.

## Discrete-time system $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right) \in \mathcal{B}(X \times U, X \times Y)$

Given **input**  $u \in \ell^2(\mathbb{N}; U)$  and **initial state**  $x_0 \in X$ , we associate the **state** trajectory  $x : \mathbb{N} \rightarrow X$  and **output**  $y : \mathbb{N} \rightarrow Y$  through

$$\begin{cases} x_{k+1} = Ax_k + Bu_k, \\ y_k = Cx_k + Du_k, \end{cases} \quad k \in \mathbb{N}. \quad (4)$$

The **transfer function**  $P(z) := D + C(z^{-1} - A)^{-1}B$  of  $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  is proper.

We call  $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  a **realization** of  $P$ .

The **Z-transform**  $\hat{u}$  of  $u : \mathbb{N} \rightarrow U$  is defined by  $\hat{u}(z) := \sum_n z^n u_n$ .

For  $x_0 = 0$ , we have  $\hat{y} = P\hat{u}$ .

## State feedback $u_k = Fx_k$

**State feedback** means that we feed the state back to the input through some **state-feedback operator**  $F \in \mathcal{B}(X, U)$ :

$$u_k := Fx_k + (u_{\text{in}})_k \quad (k \in \mathbb{N}), \quad (5)$$

where  $u_{\text{in}}$  denotes an exogenous input (or disturbance), as in Figure 3.

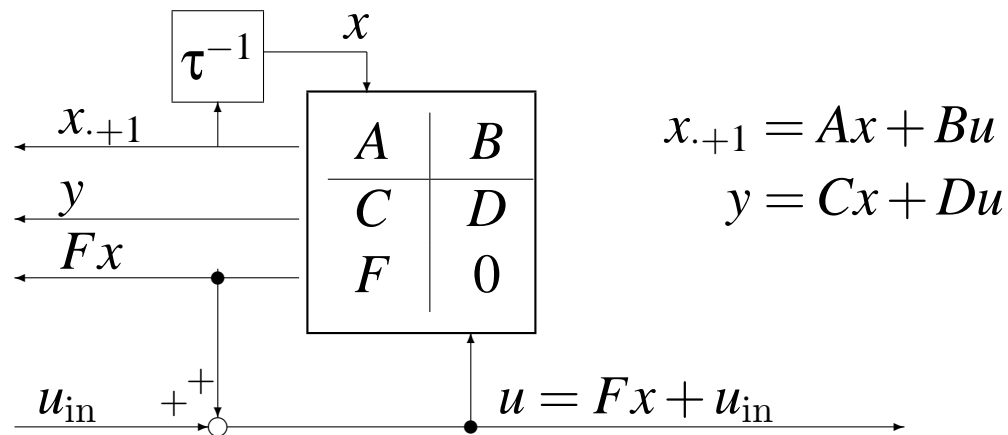


Figure 3: State-feedback connection

$$\Rightarrow x_{k+1} = (A + BF)x_k + B(u_{\text{in}})_k \Rightarrow \left( \begin{array}{c|c} A+BF & B \\ \hline C & D \\ F & I \end{array} \right) : \begin{bmatrix} x_k \\ (u_{\text{in}})_k \end{bmatrix} \mapsto \begin{bmatrix} x_{k+1} \\ y_k \\ u_k \end{bmatrix}.$$

## Closed-loop system

$$\left( \begin{array}{c|c} A + BF & B \\ \hline \begin{bmatrix} C + DF \\ F \end{bmatrix} & \begin{bmatrix} D \\ I \end{bmatrix} \end{array} \right) : \begin{bmatrix} x_k \\ (u_{\text{in}})_k \end{bmatrix} \mapsto \begin{bmatrix} x_{k+1} \\ y_k \\ u_k \end{bmatrix}. \quad (6)$$

The transfer function of the **closed-loop system** (6) is obviously given by

$$\begin{bmatrix} N(z) \\ M(z) \end{bmatrix} = \begin{bmatrix} D \\ I \end{bmatrix} + \begin{bmatrix} C + DF \\ F \end{bmatrix} (z^{-1} - A - BF)^{-1} B. \quad (7)$$

Because  $\begin{bmatrix} N \\ M \end{bmatrix}$  maps  $\widehat{u}_{\text{in}} \mapsto \begin{bmatrix} \widehat{y} \\ \widehat{u} \end{bmatrix}$ , a factorization of  $P : \widehat{u} \mapsto \widehat{y}$  is given by  $P = NM^{-1}$ .

**Finite Cost Condition (FCC):** For each  $x_0 \in \mathbb{X}$ , some  $u \in \ell^2$  makes  $y \in \ell^2$ .

If(f) the FCC holds, then there exists  $F \in \mathcal{B}(\mathbb{X}, \mathbb{U})$  that minimizes  $\sum_{k=0}^{\infty} (\|y_k\|_Y^2 + \|u_k\|_U^2)$  (LQR cost) for every  $x_0$ .

The resulting factorization  $P = NM^{-1}$  is **weakly coprime** [M05a].

If the FCC holds for  $\left( \begin{array}{c|c} A^* & C^* \\ \hline B^* & D^* \end{array} \right)$ , then  $P = NM^{-1}$  is **right coprime** [CO05].

## State-feedback stabilization of $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$

$\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  **output stable** =  $y \in \ell^2$  whenever  $x_0 \in \mathbf{X}$  and  $u = 0$ ;  
i.e.,  $\|CAx_0\|_2 \leq K\|x_0\|_{\mathbf{X}}$  ( $x_0 \in \mathbf{X}$ ).

$\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  **stable** =  $y \in \ell^2$  and  $x$  is bounded whenever  $x_0 \in \mathbf{X}$  and  $u \in \ell^2(\mathbb{N}; \mathbf{U})$ ; i.e.,

$$\|x_n\|_{\mathbf{X}} + \|y\|_2 \leq K (\|x_0\|_{\mathbf{X}} + \|u\|_2) \quad (n \geq 0, x_0 \in \mathbf{X}, u \in \ell^2(\mathbb{N}; \mathbf{U})). \quad (8)$$

$\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  **[output-]stabilizable** =  $\left(\begin{array}{c|c} A+BF & B \\ \hline C & D \end{array}\right)$  [output-]stable for some  $F$ .

$\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  **[input-]detectable** =  $\left(\begin{array}{c|c} A^* & C^* \\ \hline B^* & D^* \end{array}\right)$  [output-]stabilizable.

**(iii)**  $P$  has a **stabilizable and detectable realization**.

**(iii')**  $P$  has an **output-stabilizable and input-detectable realization**.

**Theorem** Output-stabilizability  $\Leftrightarrow$  Finite Cost Condition. [M05a]

**(iii'')**  $P$  has a realization  $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  such that  $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  and  $\left(\begin{array}{c|c} A^* & C^* \\ \hline B^* & D^* \end{array}\right)$  satisfy the **Finite Cost Condition**.

## Main theorem (ver. 4)

The following are equivalent for a proper function  $P$ :

(i)  $P$  has a proper **stabilizing controller**  $Q$  (i.e.,  $\begin{bmatrix} I & -Q \\ -P & I \end{bmatrix}^{-1} \in \mathbf{H}^\infty$ ).

(i''')  $P$  has a realization that has a **stabilizing controller system**.

(ii)  $P$  has a **right coprime** factorization  $P = NM^{-1}$ .

(iii)  $P$  has a **stabilizable and detectable realization**.

(iii')  $P$  has an **output-stabilizable and input-detectable realization**.

(iii'')  $P$  has a realization  $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  such that  $\left(\begin{array}{c|c} A & B \\ \hline C & D \end{array}\right)$  and  $\left(\begin{array}{c|c} A^* & C^* \\ \hline B^* & D^* \end{array}\right)$  satisfy the **Finite Cost Condition**.

(iii''')  $P$  has a **strongly** stabilizable and strongly detectable realization.

(“Strongly” means that, in addition,  $x_k \rightarrow 0$ , as  $k \rightarrow +\infty$ .)

## Dynamic output-feedback stabilization

(i''')  $P$  has a realization that has a **stabilizing controller system**  $\left(\begin{smallmatrix} \tilde{A} & \tilde{B} \\ \tilde{C} & \tilde{D} \end{smallmatrix}\right)$ .

This says that if we feed the output of  $\left(\begin{smallmatrix} A & B \\ C & D \end{smallmatrix}\right)$  through  $\left(\begin{smallmatrix} \tilde{A} & \tilde{B} \\ \tilde{C} & \tilde{D} \end{smallmatrix}\right)$  back to the input, then the combined system becomes stable, i.e., in Figure 4

$$\| \begin{bmatrix} x_n \\ \tilde{x}_n \end{bmatrix} \|_{\mathbf{X} \times \tilde{\mathbf{X}}} + \| \begin{bmatrix} y \\ u \end{bmatrix} \|_2 \leq K \left( \| \begin{bmatrix} x_0 \\ \tilde{x}_0 \end{bmatrix} \|_{\mathbf{X} \times \tilde{\mathbf{X}}} + \| \begin{bmatrix} y_{\text{in}} \\ u_{\text{in}} \end{bmatrix} \|_2 \right) \quad (n \geq 0, \begin{bmatrix} x_0 \\ \tilde{x}_0 \end{bmatrix} \in \mathbf{X} \times \tilde{\mathbf{X}}, \begin{bmatrix} y_{\text{in}} \\ u_{\text{in}} \end{bmatrix} \in \ell^2(\mathbb{N}; \mathbf{Y} \times \mathbf{U})).$$

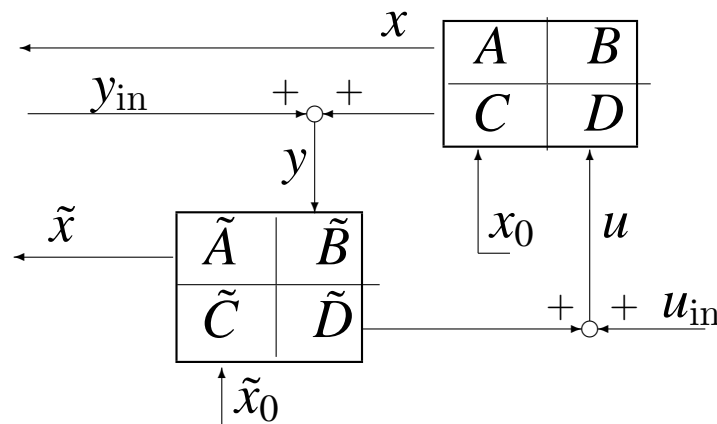


Figure 4: Stabilizing controller system

This implies that  $Q(z) = \tilde{D} + \tilde{C}(z^{-1} - \tilde{A})^{-1}\tilde{B}$  is a proper stabilizing controller for  $P(z) = D + C(z^{-1} - A)^{-1}B$  (i.e.,  $\begin{bmatrix} I & -Q \\ -P & I \end{bmatrix}^{-1} \in H^\infty$ ).

The converse holds iff  $\left(\begin{smallmatrix} A & B \\ C & D \end{smallmatrix}\right)$  and  $\left(\begin{smallmatrix} \tilde{A} & \tilde{B} \\ \tilde{C} & \tilde{D} \end{smallmatrix}\right)$  are stabilizable and detectable [M05e].

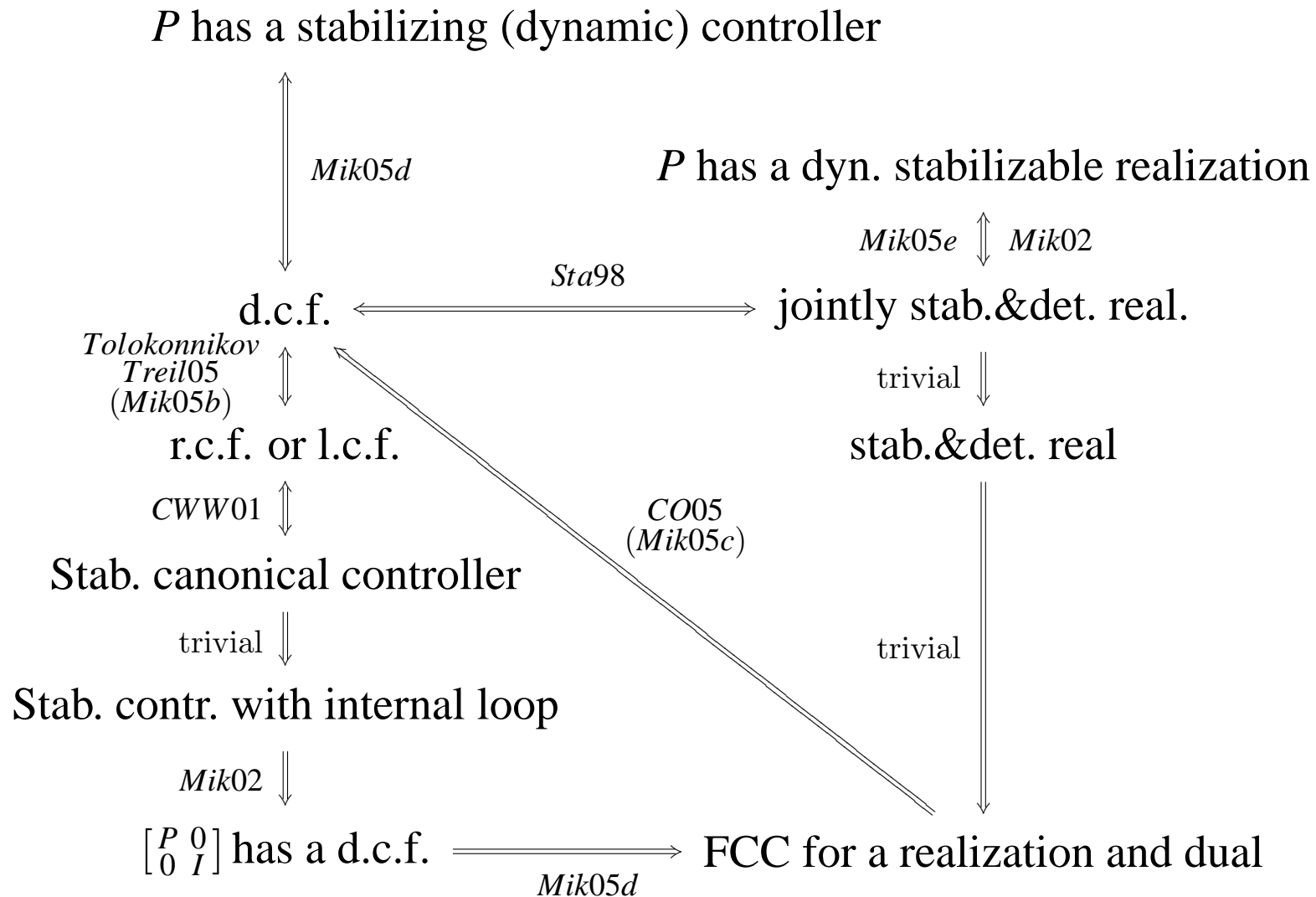
## Main theorem (final version)

The following are equivalent for a proper function  $P$ :

- (i)  $P$  has a proper **stabilizing controller**  $Q$  (i.e.,  $\begin{bmatrix} I & -Q \\ -P & I \end{bmatrix}^{-1} \in \mathbf{H}^\infty$ ).
- (i'')  $P$  has a stabilizing controller with **internal loop**. [CuWeWe01] [M05d]
- (i''')  $P$  has a realization that has a **stabilizing controller system**  $\left(\begin{array}{c|c} \tilde{A} & \tilde{B} \\ \hline \tilde{C} & \tilde{D} \end{array}\right)$ . [M05e]
- (ii)  $P$  has a **right coprime** factorization  $P = NM^{-1}$ . [Smith] [M05d]
- (ii'')  $P$  has a **doubly coprime** factorization  $P = NM^{-1}$ ,  $\begin{bmatrix} M & Y \\ N & X \end{bmatrix}, \begin{bmatrix} M & Y \\ N & X \end{bmatrix}^{-1} \in \mathbf{H}^\infty(\mathbf{U} \times \mathbf{Y})$ .
- (iii)  $P$  has a **stabilizable and detectable realization**. [St98] [CO05] [M05c]



# Proof of part of the Main Theorem



## Weaker equivalent conditions

- (i)  $P$  has a proper **stabilizing controller**  $Q$  (i.e.,  $\begin{bmatrix} I & -Q \\ -P & I \end{bmatrix}^{-1} \in H^\infty$ ).
- (ii)  $P$  has a **right coprime** factorization  $P = NM^{-1}$ .
- (ii'')  $P$  has a **doubly coprime** factorization  $P = NM^{-1}$ ,  $\begin{bmatrix} M & Y \\ N & X \end{bmatrix}, \begin{bmatrix} M & Y \\ N & X \end{bmatrix}^{-1} \in H^\infty(U \times Y)$ .
- (iii)  $P$  has a **stabilizable and detectable** realization.
- (iii')  $P$  has an **output-stabilizable** realization whose **dual** is output-stabilizable.

The following (strictly weaker) conditions equivalent to each other:

- (ii-)  $P$  has a **factorization**  $P = NM^{-1}$  ( $N, M \in H^\infty$ ).
- (ii'-)  $P$  has a **weakly coprime** factorization  $P = NM^{-1}$ . [M05a]
- (iii-)  $P$  has a **stabilizable realization**.
- (iii'-)  $P$  has an output-stabilizable realization. [M02]
- (i-) The **range** of the generalized **Hankel** operator of  $P$  lies in the **range** of the generalized **Toeplitz** operator of  $P$  plus  $H^2$ . [M05c]

## Weakly coprime = common right factors are units

Scalar-valued case ( $U = \mathbb{C} = Y$ ):

$N \in H^\infty(U, Y)$  and  $M \in H^\infty(U, U)$  are weakly coprime iff  $\gcd(N, M) = I$ .

Equivalent condition: if  $N = N_1V$  and  $M = M_1V$ , then  $V$  is a unit ( $V, V^{-1} \in H^\infty$ ).

[Fuhrmann81] [Smith89]

Equivalent condition:

if  $N = N_1V$  and  $M = M_1V$  and  $V(0)$  is invertible, then  $V$  is a unit ( $V, V^{-1} \in H^\infty$ ).

I.e., every properly-invertible common right factor is a unit.

This latter condition is meaningful also when  $U$  and  $Y$  are infinite-dimensional (the former is then never satisfied). An equivalent condition is

$$Nf, Mf \in H^2 \implies f \in H^2 \quad (9)$$

(for every proper  $U$ -valued function  $f$ ). Either of these two conditions can be used as the definition of **weak right coprimeness**.

One obtains a third equivalent condition by replacing  $H^2$  by  $H^\infty$  in (9).

# Generalized Toeplitz and Hankel ranges

- (ii-)  $P$  has a **factorization**  $P = NM^{-1}$  ( $N, M \in H^\infty$ ).
- (ii'-)  $P$  has a **weakly coprime** factorization  $P = NM^{-1}$ .
- (iii-)  $P$  has a **stabilizable realization**.
- (i-)  $\text{Ran}(H_P) \subset \text{Ran}(T_P) + H^2$ .
- (i'-)  $\exists r > 1 \forall v \in \ell_r^2(\mathbb{Z}_-; U) \exists u \in \ell^2(\mathbb{N}; U)$  such that  $\mathcal{D}(v+u) \in \ell^2(\mathbb{N}; Y)$

$T_P$  is the “unbounded Toeplitz operator” that maps  $H^2 \ni \hat{u} \mapsto P\hat{u}$

$H_P$  is the “unbounded Hankel operator” that maps  $H^2(r\mathbb{D}^-; U) \ni \hat{v} \mapsto$  projection of  $P\hat{v}$  onto  $H_r^2 := H^2(r\mathbb{D}; Y)$  (for some big  $r$ ).

The  $I/O$  map  $\mathcal{D}$  is determined by  $\widehat{\mathcal{D}u} = P\hat{u}$ . It has a unique continuous extension to a map  $\mathcal{D} : \ell_r^2 \rightarrow \ell_r^2$  for every big  $r$ , where  $\|u\|_{\ell_r^2}^2 := \sum_{k=-\infty}^{\infty} r^{2k} \|u_k\|_U^2$ .

Note that  $T_P\hat{u} = \widehat{(\pi_+ \mathcal{D} \pi_+)}$  and  $H_P\hat{u} = \widehat{(\pi_+ \mathcal{D} \pi_-)}$ ,  
 where  $(\pi_+ u)_k := \begin{cases} u_k, & k \geq 0; \\ 0, & k < 0 \end{cases}$ ,  $\pi_- := I - \pi_+$ . We have set  $r\mathbb{D}^- := \{z \in \mathbb{C} \mid |z| > r\}$ .

Naturally,  $\|\hat{u}\|_{H_r^2}^2 := \|\hat{u}(r \cdot)\|_{H^2}^2 = \sup_{t < r} \int_0^{2\pi} \|\hat{u}(te^{i\theta})\|_U^2 d\theta = 2\pi \|\hat{u}\|_{\ell_r^2}^2$ .

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